



# Unfolding

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# Introduction (1/4)

- **Unfolding** is a transformation technique that can be applied to a DSP program to create a new program describing more than one iterations of the original program
- Unfolding factor  $J$ :  $J$  consecutive iterations
- Also called as loop unrolling



# Introduction (2/4)

- For the DSP algorithm

*for  $n = 0 \rightarrow \infty$*

- $y(n) = ay(n-9) + x(n)$

- Replace  $n$  with  $2k$  and  $2k+1$

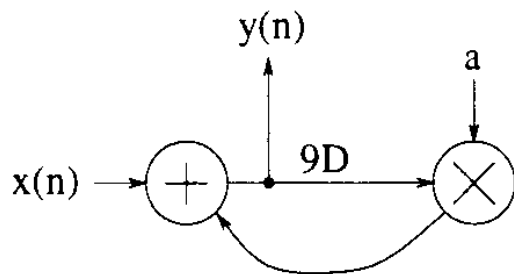
*for  $k = 0 \rightarrow \infty$*

- $y(2k) = ay(2k-9) + x(2k)$

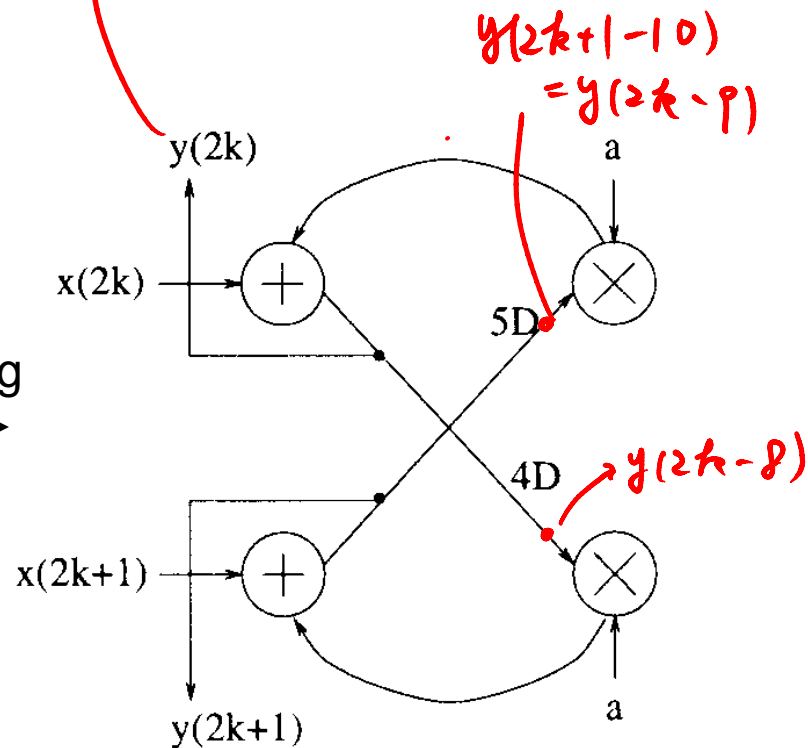
- $y(2k+1) = ay(2k-8) + x(2k+1)$

- It is an unfolded algorithm with  $J=2!$

# Introduction (3/4)



Unfolding  $\longrightarrow$



- Note that, in unfolded systems, each delay is J-slow

!

$$y(2k+1) = a \cdot y(2k-8) + x(2k+1)$$



# Introduction (4/4)

- Applications of unfolding
  - To reveal hidden concurrent so that the program can be scheduled to a smaller iteration period
  - To design parallel architecture

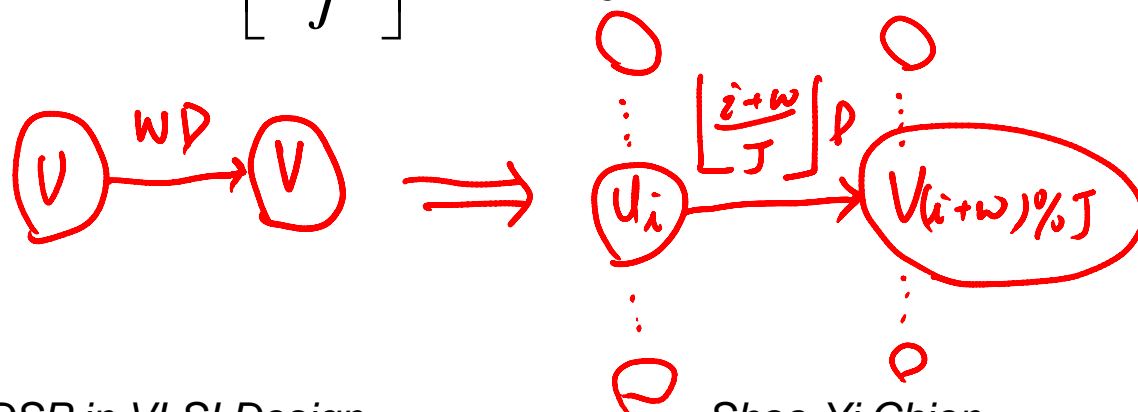


# Algorithm for Unfolding

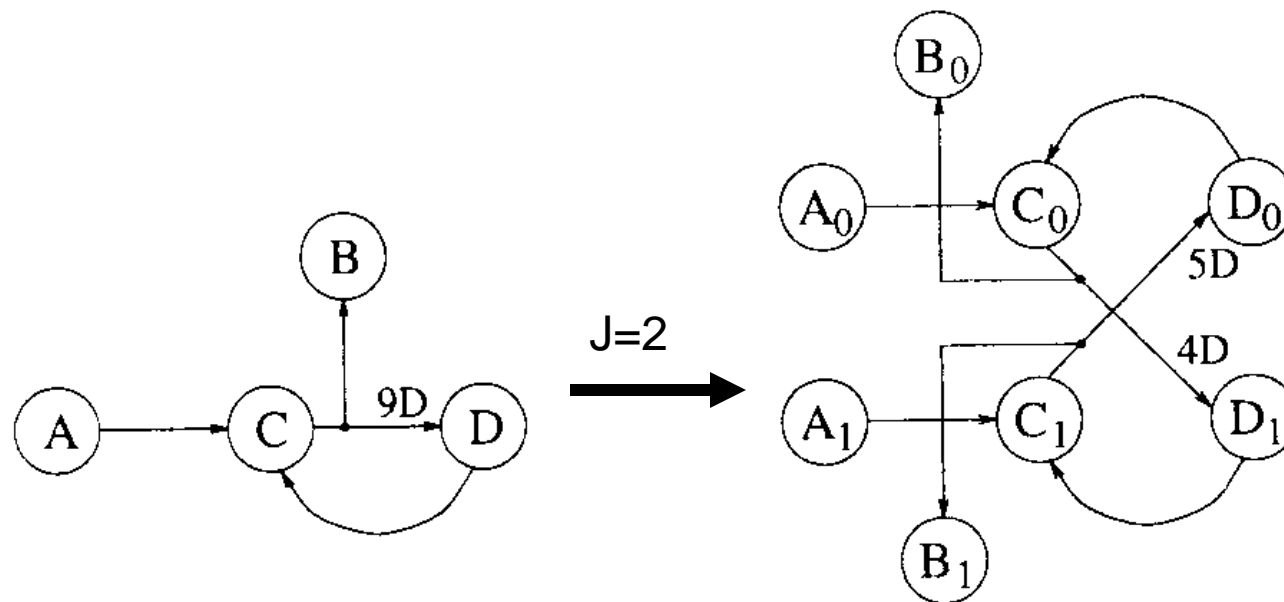
- In the J-unfolded DFG
  - For each node U in the origin DFG, there are J nodes with the same function as U
  - For each edge in the original DFG, there are J edges

# Algorithm for Unfolding

- For each node  $U$  in the original DFG, draw the  $J$  nodes  $U_0, U_1, \dots, U_{J-1}$
- For each edge  $U \rightarrow V$  with  $w$  delays in the original DFG, draw the  $J$  edges  $U_i \rightarrow V_{(i+w)\%J}$  with  $\left\lfloor \frac{i+w}{J} \right\rfloor$  delays for  $i=0, 1, \dots, J-1$



# Example 1 of Unfolding

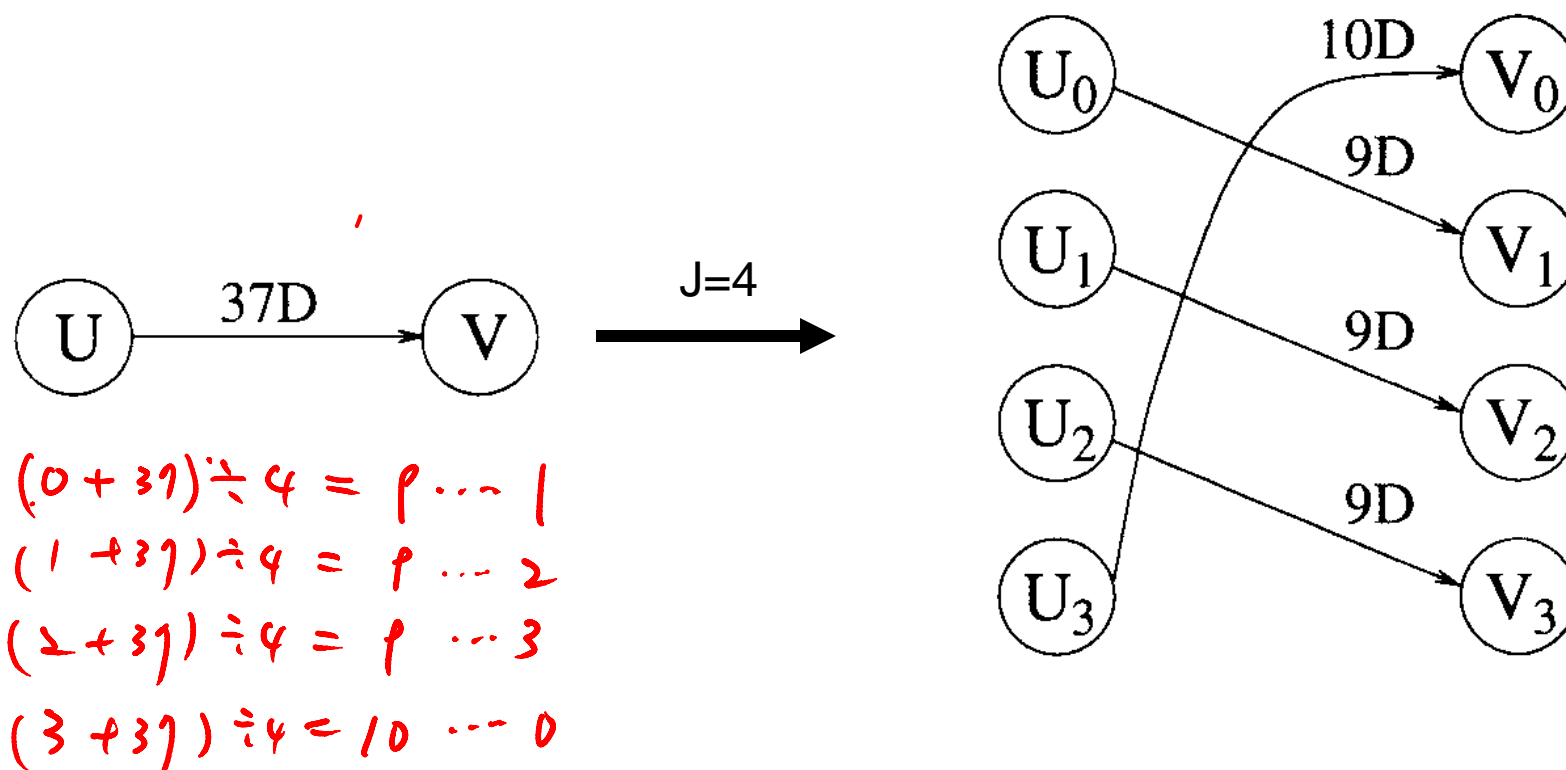


$$C_0 \rightarrow D_{(0+9)\%2} \text{ with } \left\lfloor \frac{0+9}{2} \right\rfloor \text{ delays}$$

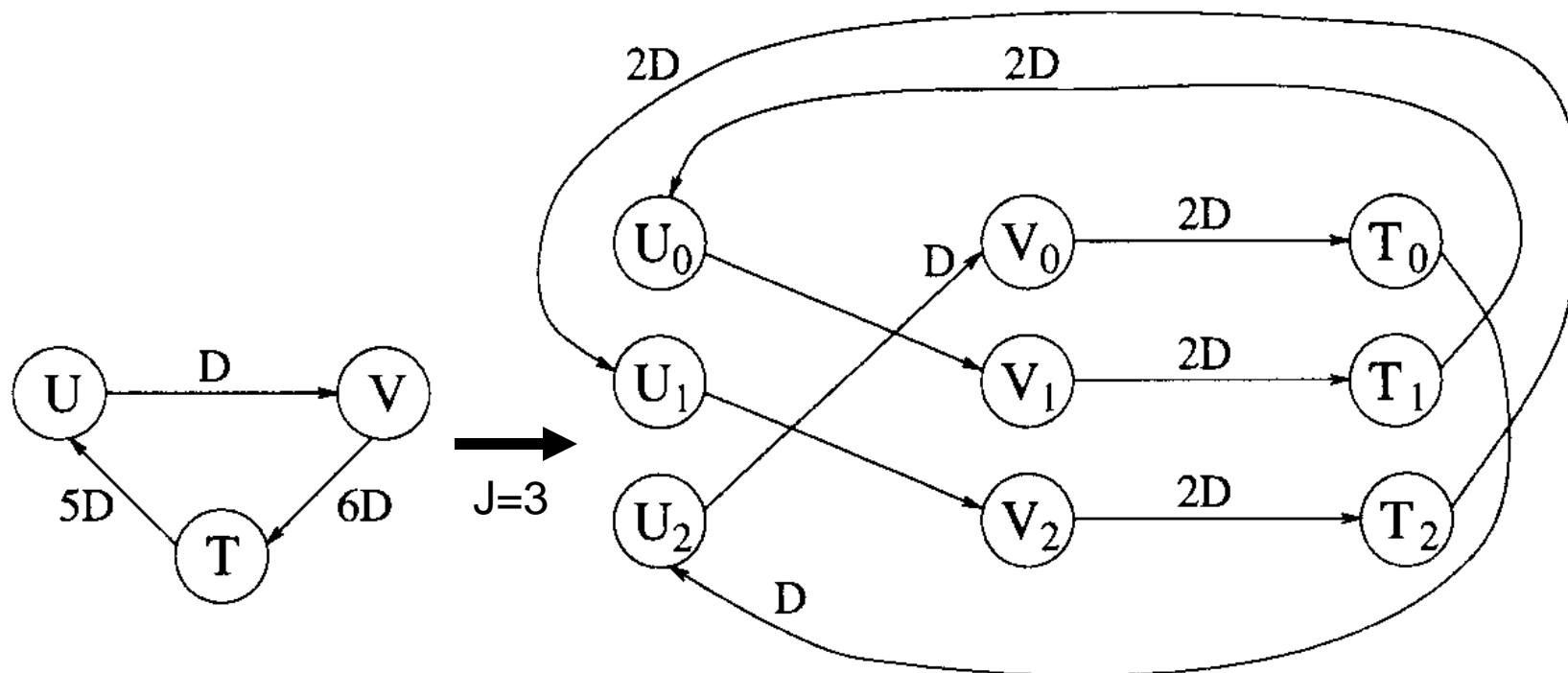
$$C_1 \rightarrow D_{(1+9)\%2} \text{ with } \left\lfloor \frac{1+9}{2} \right\rfloor \text{ delays}$$



# Example 2 of Unfolding



# Example 3 of Unfolding



# Proof of the Unfolding Algorithm (1/2)

- Unfolding preserve precedence constraints of a DSP program

- For  $U_i \rightarrow V_{(i+w)\%J}$  with  $\left\lfloor \frac{i+w}{J} \right\rfloor$  delays

output of  $U_i$  in the  $k$ -th iteration will be connected to  $V_{(i+w)\%J}$  in the  $(k + \left\lfloor \frac{i+w}{J} \right\rfloor)$ -th iteration

- In the original DFG, it corresponds to:

output of  $U$  in the  $(Jk+i)$ -th iteration will be connected to  $V$  in the  $(J(k + \left\lfloor \frac{i+w}{J} \right\rfloor) + (i+w)\%J)$ -th iteration

# Proof of the Unfolding Algorithm (2/2)

$$\begin{aligned} & J \left( k + \left\lfloor \frac{i+w}{J} \right\rfloor \right) + (i+w) \% J - (Jk + i) \\ &= \left( J \left\lfloor \frac{i+w}{J} \right\rfloor + (i+w) \% J \right) - i \\ &= \underline{(i+w)} - i = w \end{aligned}$$

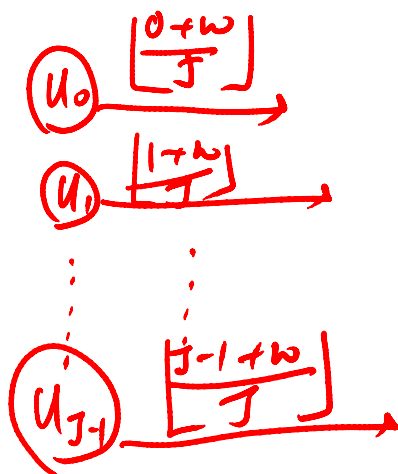
- So the precedence constraints are preserved correctly



# Properties of Unfolding (1/5)

- Unfolding preserves the number of delays in a DFG

$$\left\lfloor \frac{w}{J} \right\rfloor + \left\lfloor \frac{w+1}{J} \right\rfloor + \dots + \left\lfloor \frac{w+J-1}{J} \right\rfloor = w$$





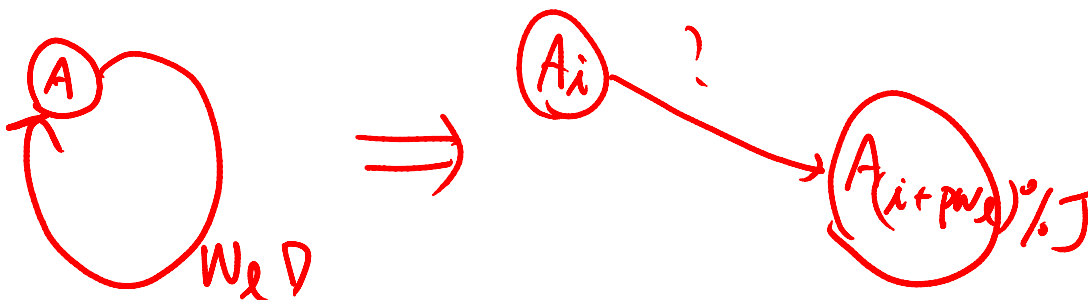
# Properties of Unfolding (2/5)

- J-unfolding of a loop  $l$  with  $w_l$  delays in the original DFG leads to  $\gcd(w_l, J)$  loops in the unfolded DFG, and each of these  $\gcd(w_l, J)$  loops contains  $w_l/\gcd(w_l, J)$  delays and  $J/\gcd(w_l, J)$  copies of each node that appears in  $l$

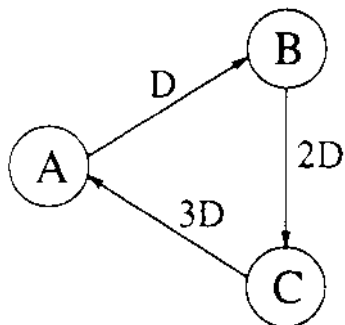


# Properties of Unfolding (3/5)

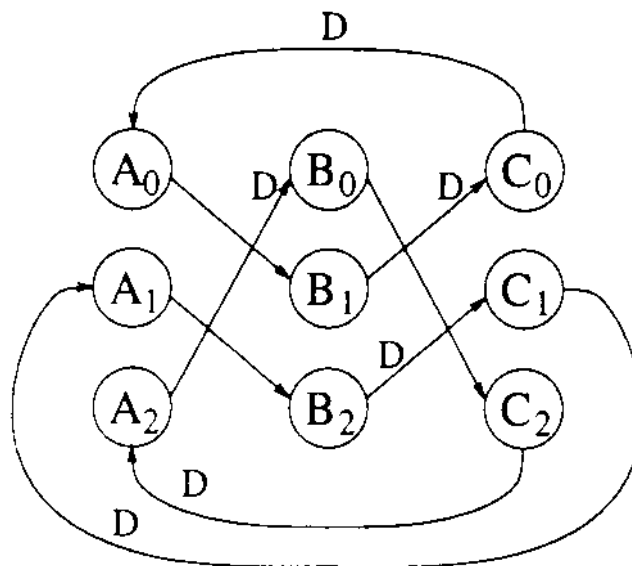
- For a loop in origin loop  $A \rightarrow A$  traversed  $p$  times with  $w_l$  delay elements
- In the unfolded DFG:  $A_i \rightarrow A_{(i+pw_l)\%J}$
- This path form a loop if  $i = (i + pw_l)\%J$



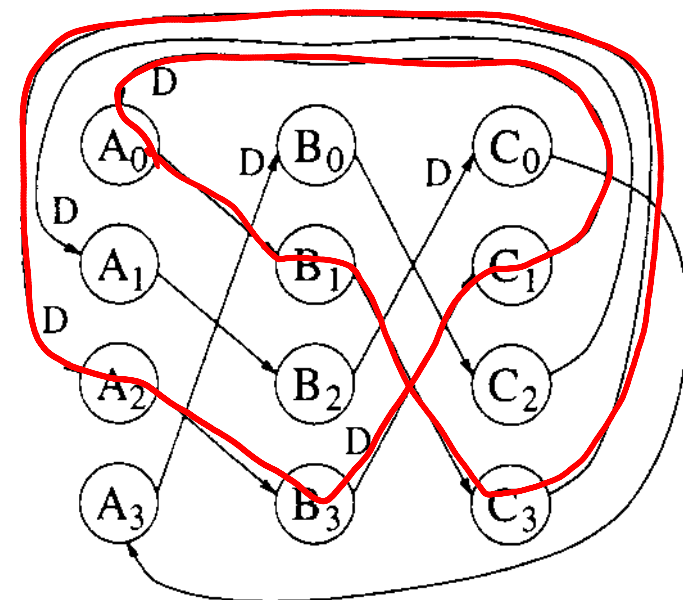
# Properties of Unfolding (4/5)



Origin DFG  
 $w_i=6$



$J=3$   
 For a loop,  $i=(i+6p)\%3$   
 $p=1$   
 Loop:  $A \rightarrow B \rightarrow C \rightarrow A$   
 Consists of 3 loops

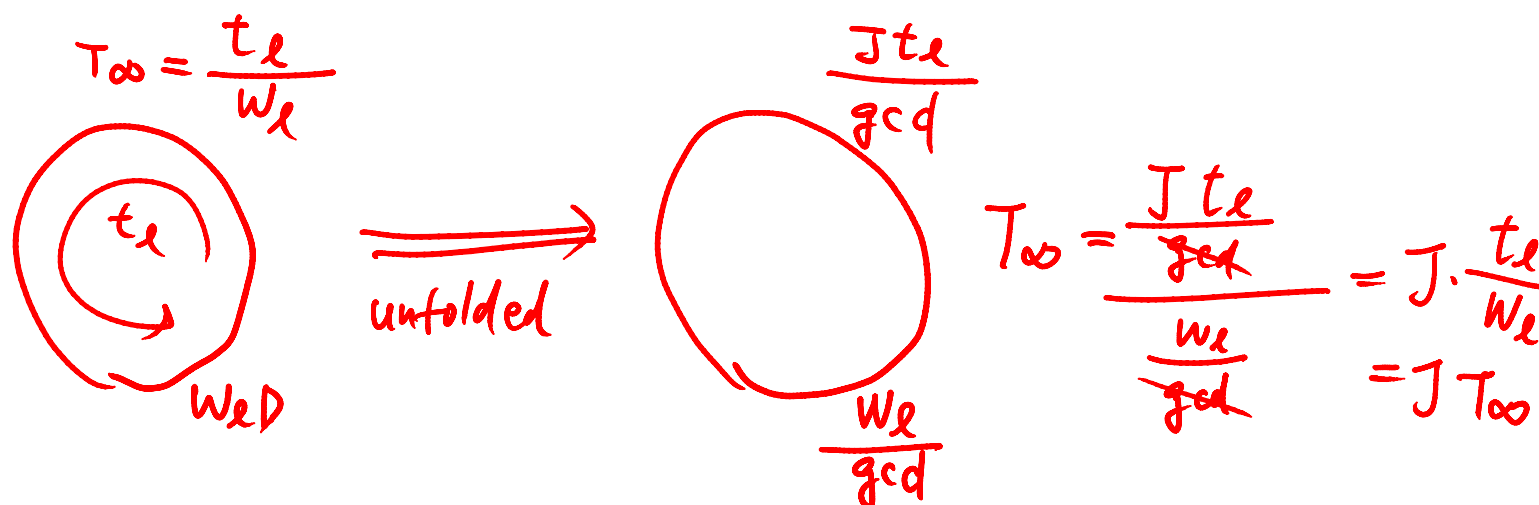


$J=4$   
 For a loop,  $i=(i+6p)\%4$   
 $p=2$   
 Loop:  $A \rightarrow B \rightarrow C \rightarrow A \rightarrow B \rightarrow C \rightarrow A$   
 Consists of 2 loops



# Properties of Unfolding (5/5)

- Unfolding a DFG with iteration bound  $T_\infty$  results in a J-unfolded DFG with iteration bound  $JT_\infty$





# Retiming with Unfolding (1/2)

- Consider a path with  $w$  delays in the original DFG.  $J$ -unfolding of this path leads to  $(J-w)$  paths with no delays and  $w$  paths with 1 delay each, when  $w < J$
- Any path in the original DFG containing  $J$  or more delays leads to  $J$  paths with 1 or more delays in each path. Therefore, a path in the original DFG with  $J$  or more delays cannot create a critical path in the  $J$ -unfolded DFG



# Retiming with Unfolding (2/2)

- The critical path of the unfolded DFG can be  $c$  if there exists a path in the original DFG with computation time  $c$  and less than  $J$  delay elements
  
- If  $D(U, V) \geq c$ ,  $W_r(U, V) = W(U, V) + r(V) - r(U) \geq J$
- $w(e) + r(V) - r(U) \geq 0$



# Applications of Unfolding

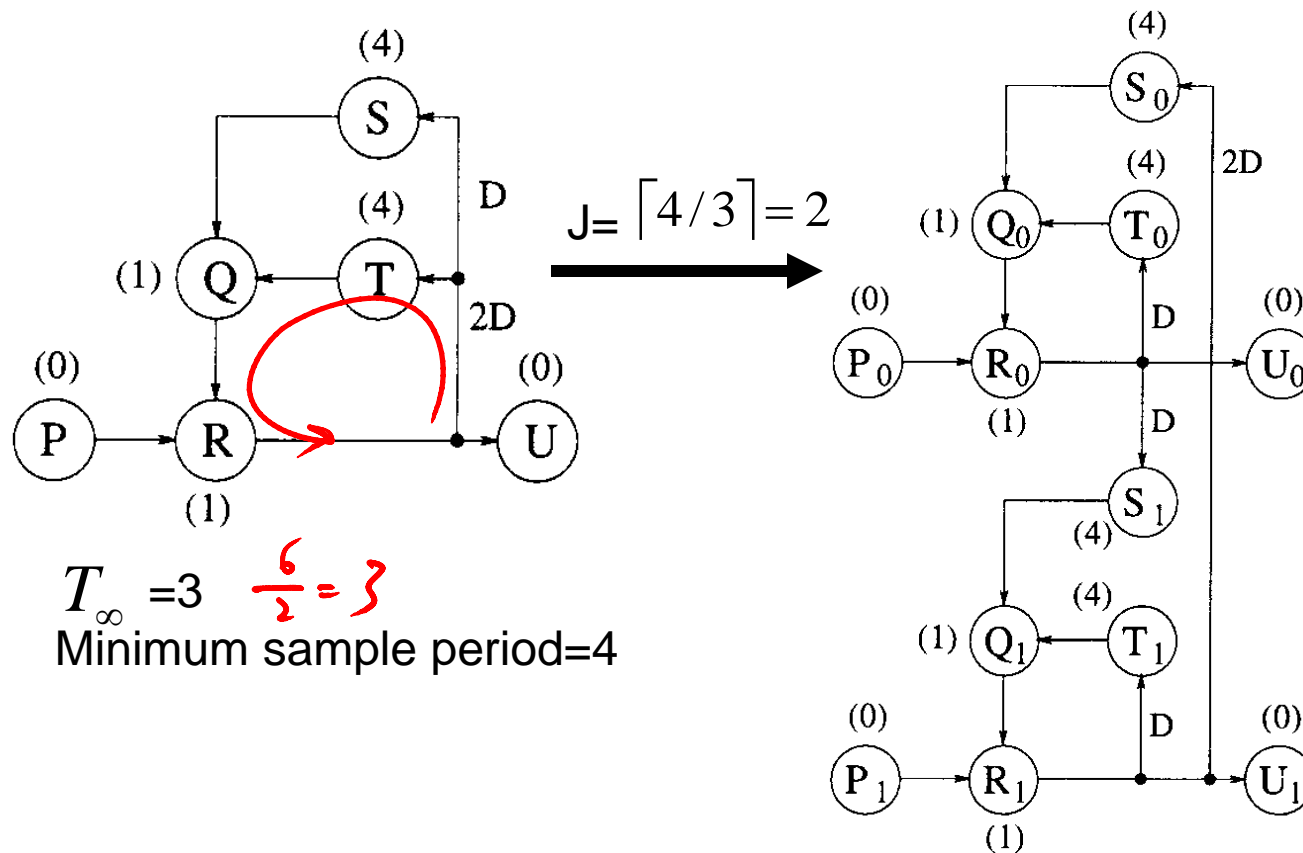
- Sample period reduction
- Parallel processing
  - Word-level parallel processing
  - Bit-level parallel processing



# Sample Period Reduction (1/5)

- In some cases, the DSP program cannot be implemented with iteration period equal to the iteration bound  $\rightarrow$  use unfolding
- First case: there is a node in the DFG that has computation time greater than  $T_\infty$ 
  - If  $t_U$  is greater than the iteration bound, then  $\lceil t_U / T_\infty \rceil$ -unfolding should be used

# Sample Period Reduction (2/5)

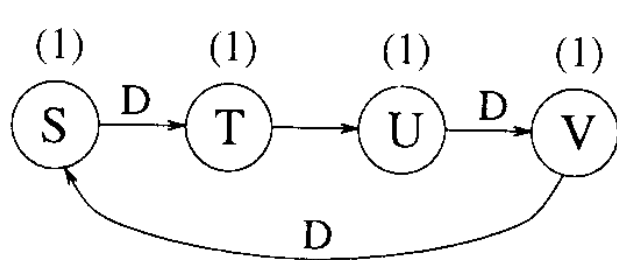




# Sample Period Reduction (3/5)

- Second case: the iteration bound is not an integer
  - If a critical loop bound is of the form  $t_1/w_1$ , where  $t_1$  and  $w_1$  are mutually coprime, then  $w_1$ -unfolding should be used

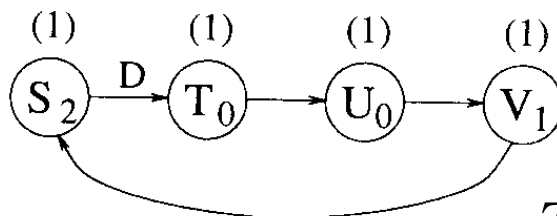
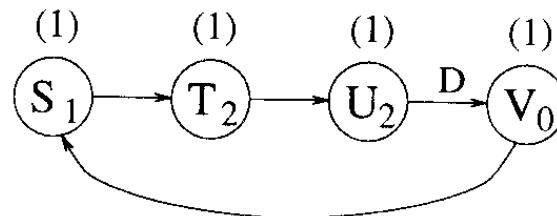
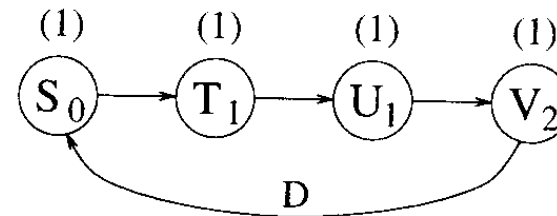
# Sample Period Reduction (4/5)



$$T_{\infty} = 4/3$$

Minimum sample period=2

$J=3$



$$T_{\infty} = 4$$

Minimum sample period=4/3





# Sample Period Reduction (5/5)

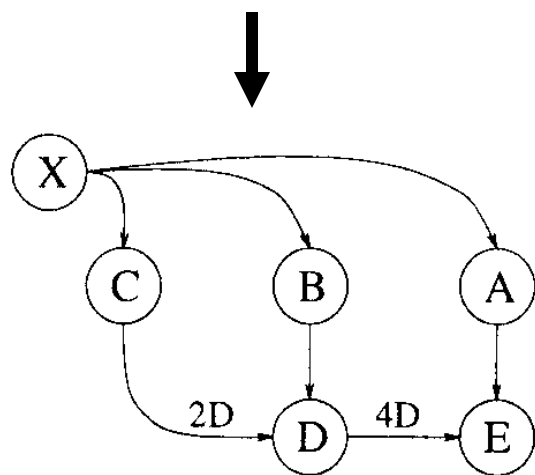
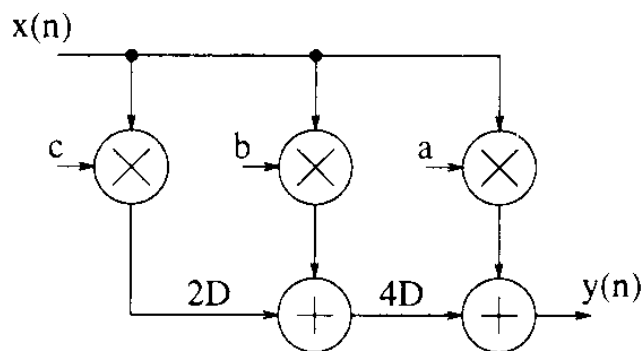
- For both cases, where the longest node computation time is larger than the iteration bound  $T_\infty$ , and  $T_\infty$  is not an integer
  - J is the minimum value such that  $JT_\infty$  is an integer and is greater than or equal to the longest node computation time



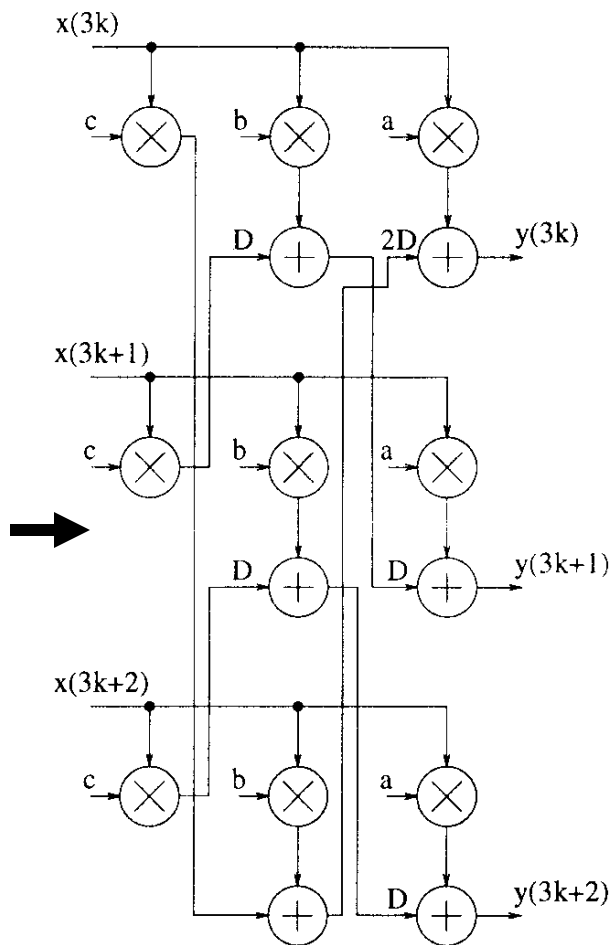
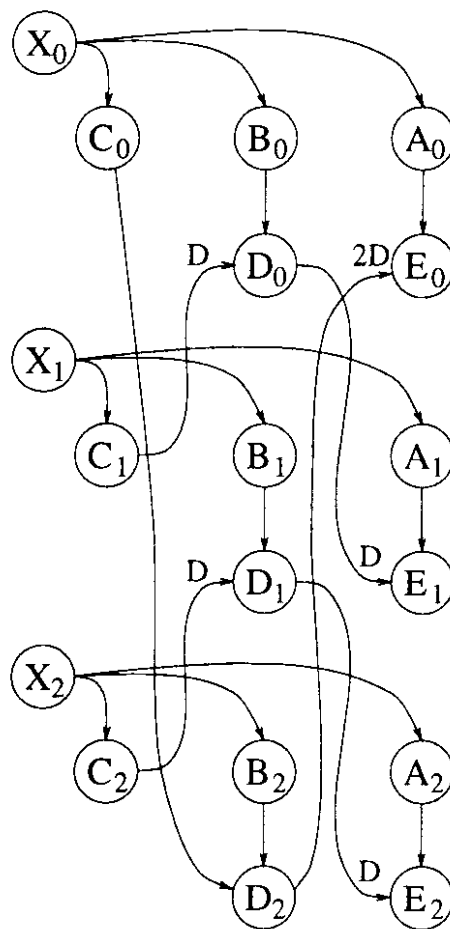
# Word-Level Parallel Processing (1/2)

- The unfolding technique can be used to design a word-parallel architecture from a word-serial architecture
  - Unfolding a word-serial architecture by  $J$  creates a word-parallel architecture that processes  $J$  words per clock cycle
  - Parallel processing

# Word-Level Parallel Processing (2/2)



$J=3$

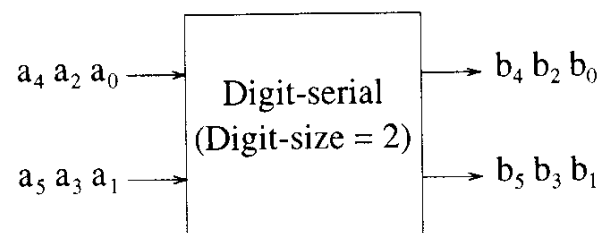
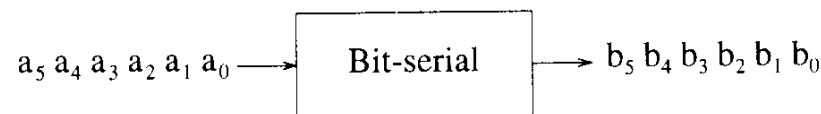
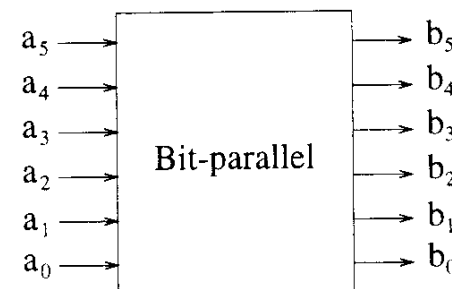


# Bit-Level Parallel Processing

## (1/6)

- Bit-parallel and bit-serial architecture can be derived from bit-serial architectures using the unfolding transformation

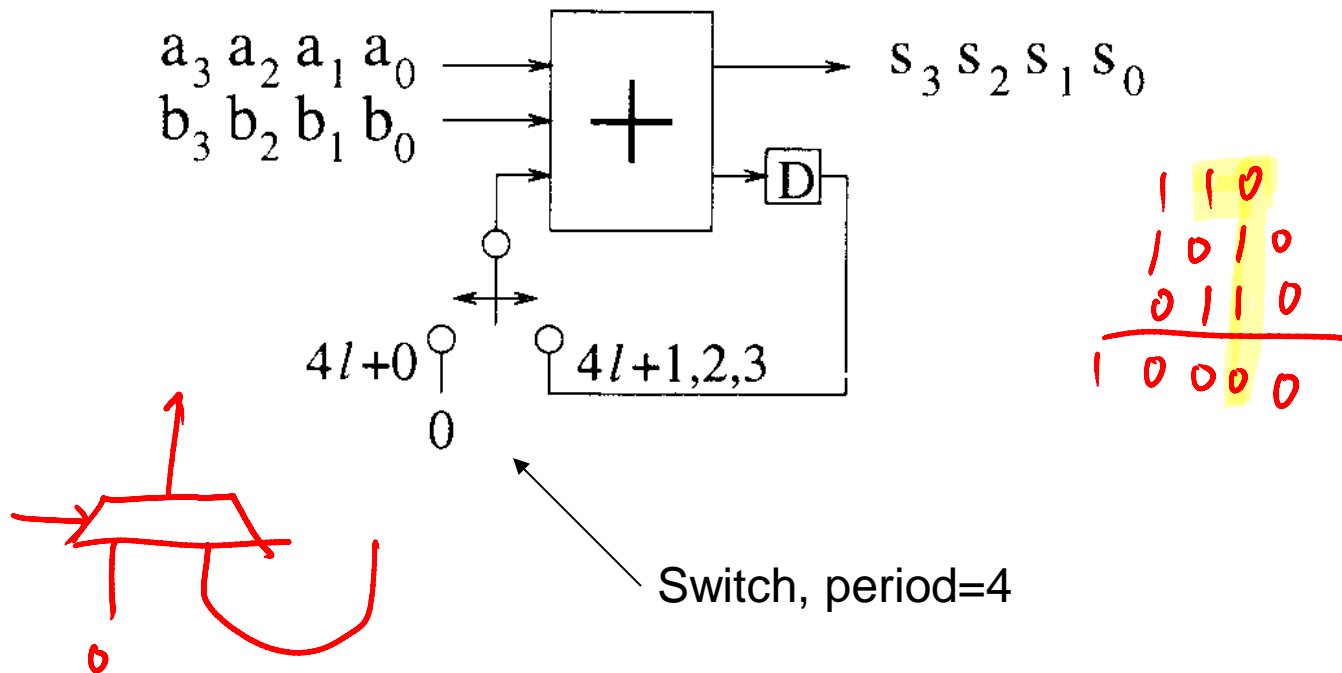
- Bit-serial
- Bit-parallel: word-length  $W$
- Digit-serial:  $N$  digits



# Bit-Level Parallel Processing

## (2/6)

### ■ Bit-serial adder for $W=4$



# Bit-Level Parallel Processing

## (3/6)

### ■ Unfolding the switch

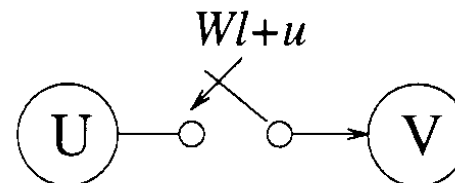
□ Assume  $W=W'J$

□ Assume all edges have no delays

□ Write the switch instance as

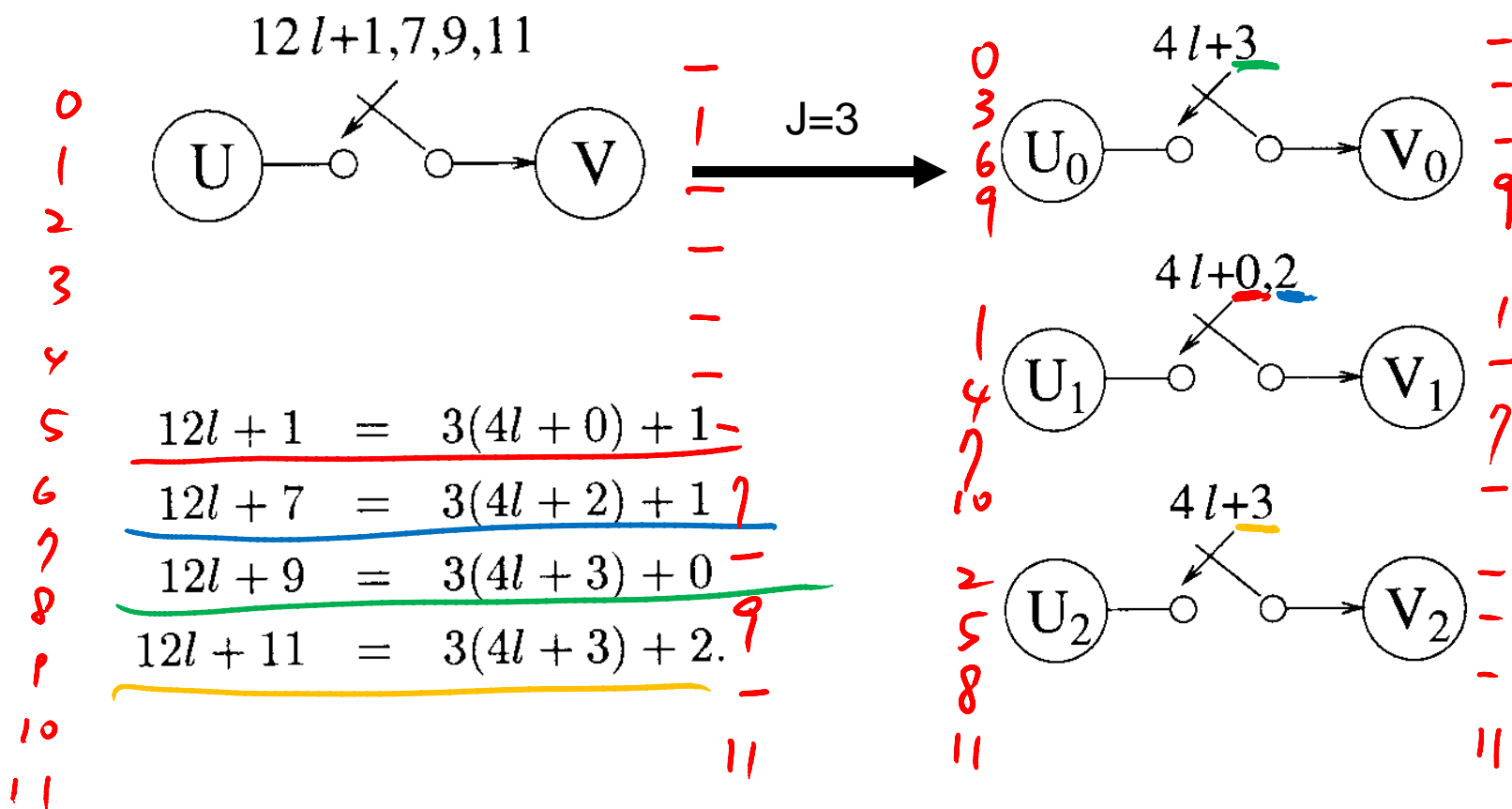
$$Wl + u = J \left( W'l + \left\lfloor \frac{u}{J} \right\rfloor \right) + (u \% J).$$

□ Draw an edge with no delays in the unfolded graph from the node  $U_{u \% J}$  to the node  $V_{u \% J}$ , which is switched at time instance  $(W'l + \left\lfloor \frac{u}{J} \right\rfloor)$





# Bit-Level Parallel Processing (4/6)



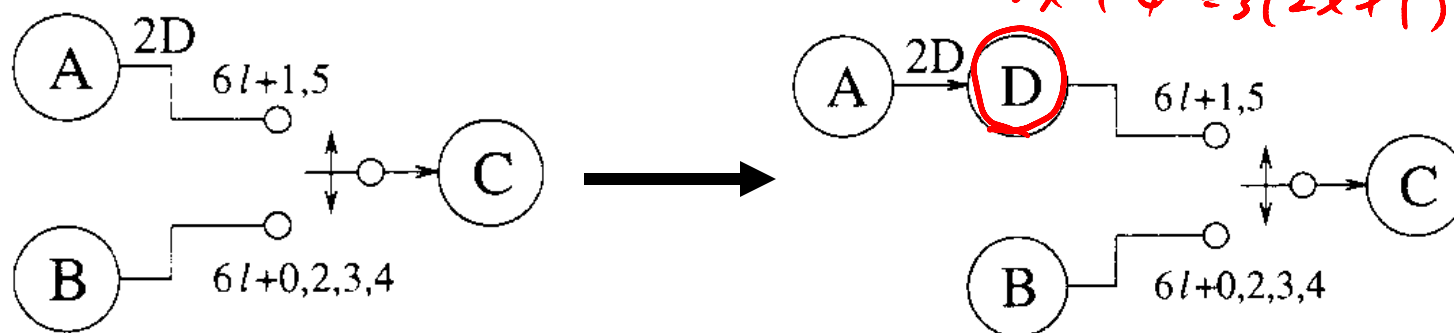
# Bit-Level Parallel Processing

## (5/6)

### ■ For edges with delays

- Add dummy nodes

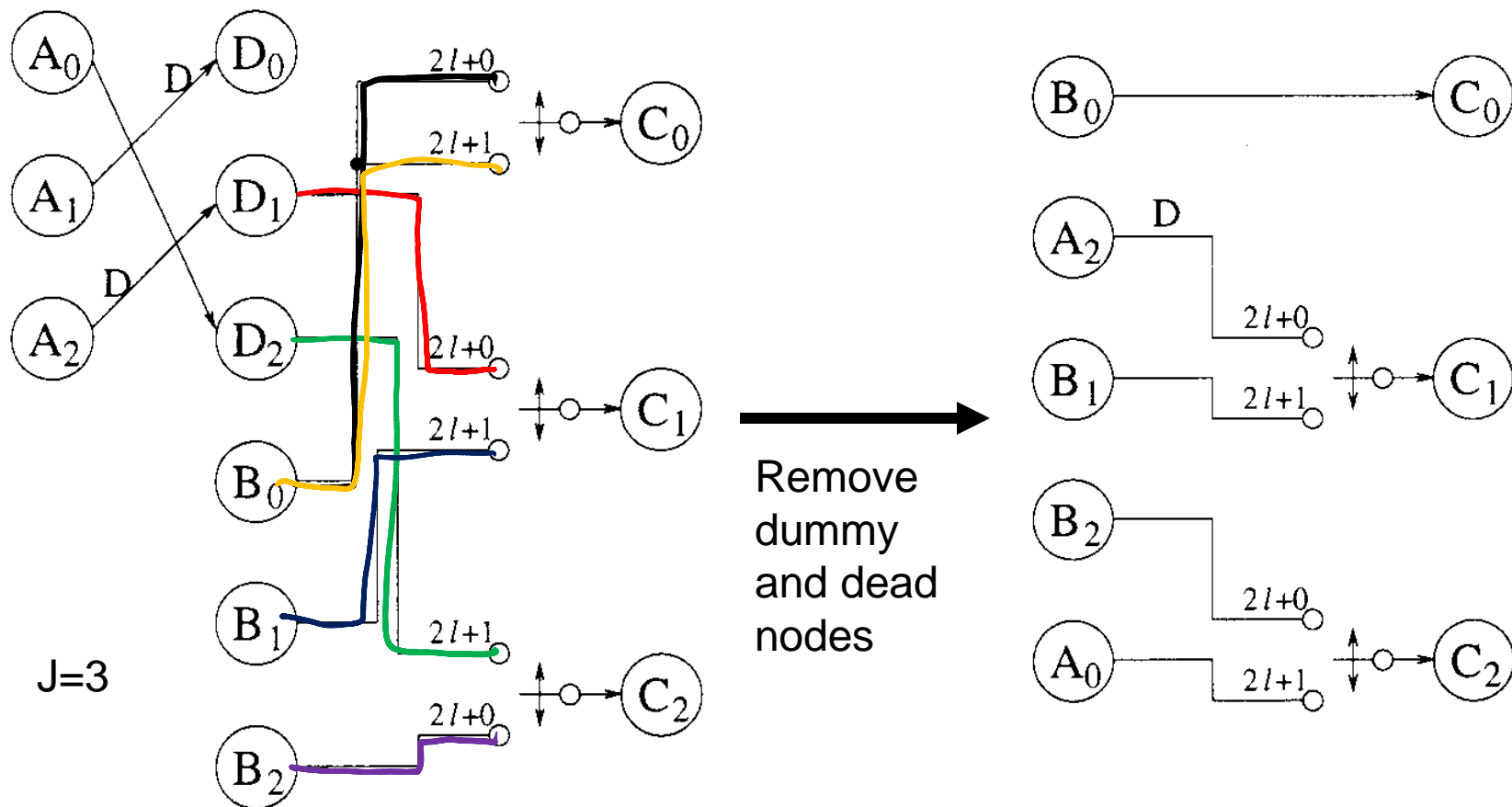
- $6l + 1 = 3(2l + 0) + 1$
- $6l + 5 = 3(2l + 1) + 2$
- $6l + 0 = 3(2l + 0) + 0$
- $6l + 2 = 3(2l + 0) + 2$
- $6l + 3 = 3(2l + 1) + 0$
- $6l + 4 = 3(2l + 1) + 1$





# Bit-Level Parallel Processing

## (6/6)





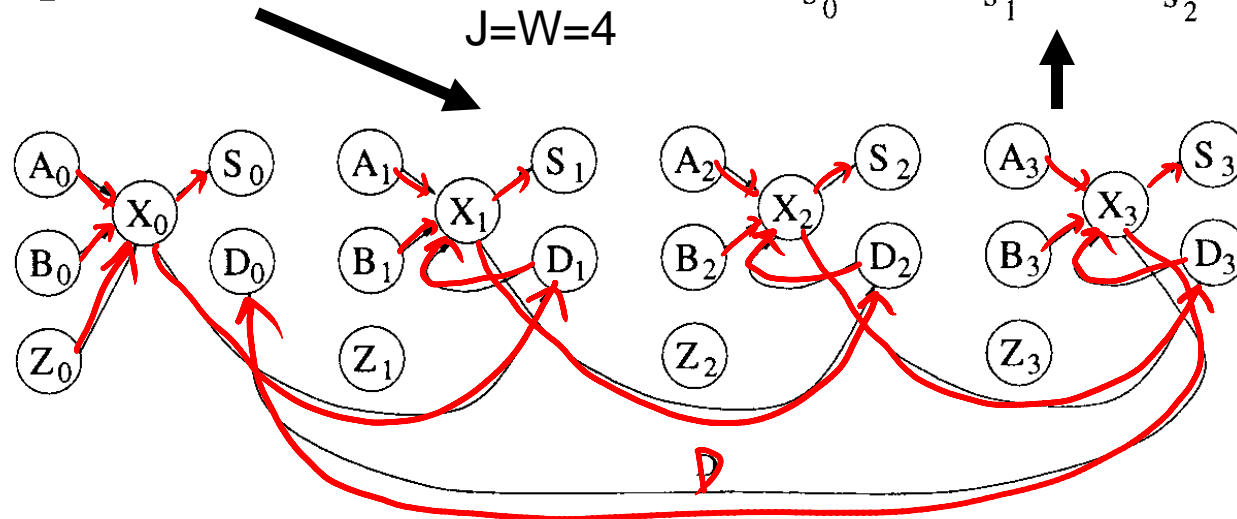
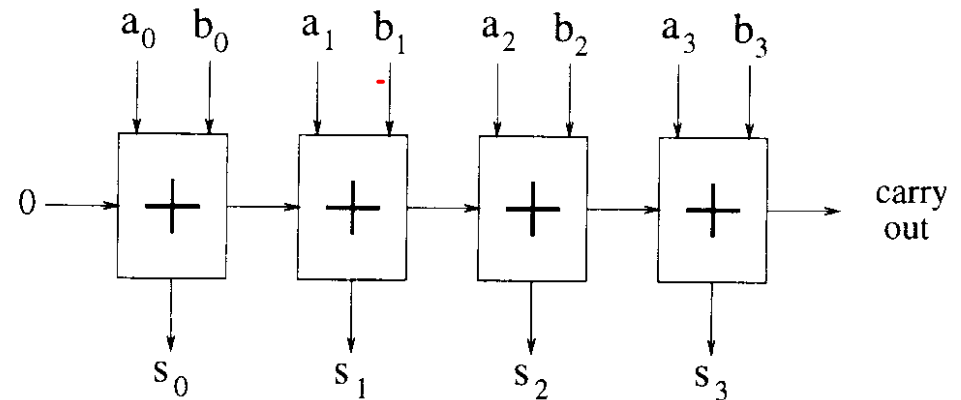
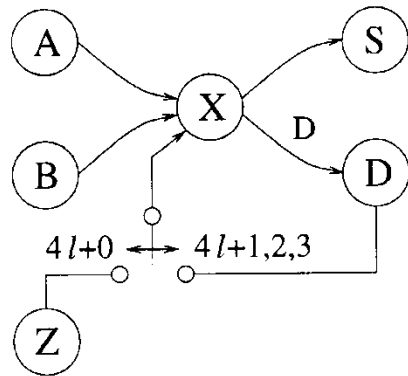
$$(4l+0) \div 4 = l \dots 0$$

$$(4l+1) \div 4 = l \dots 1$$

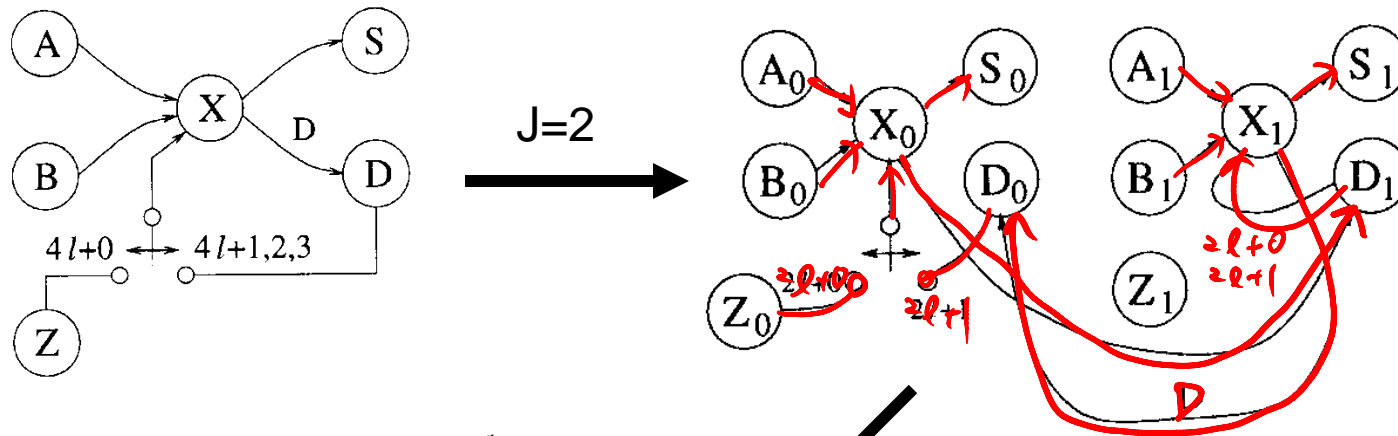
$$(4l+2) \div 4 = l \dots 2$$

$$(4l+3) \div 4 = l \dots 3$$

# Bit-Parallel Adder

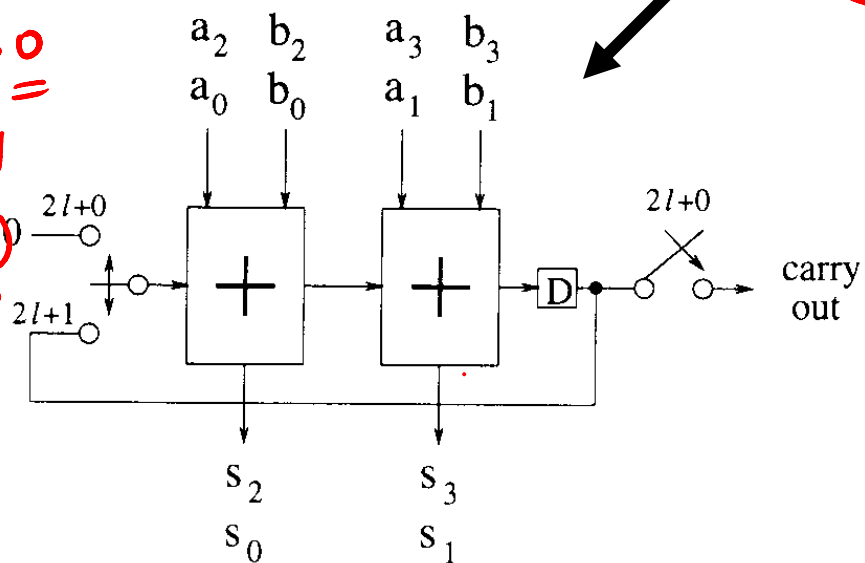


# Digit-Serial Adder (1/4)



$J=2$

$(4l+0) \div 2 = 2l \dots 0$   
 $(4l+1) \div 2 = 2l \dots 1$   
 $(4l+2) \div 2 = (2l+1) \dots 0$   
 $(4l+3) \div 2 = (2l+1) \dots 1$

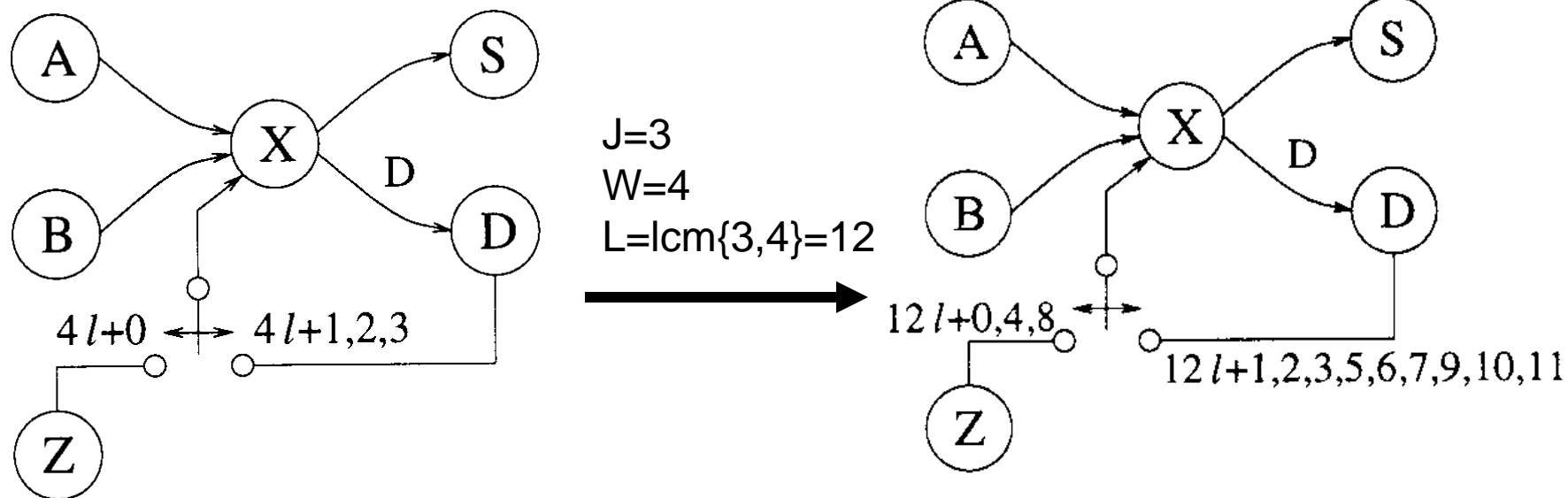




# Digit-Serial Adder (2/4)

- If  $W$  is not a multiple of the unfolding factor  $J$ 
  - $L = \text{lcm}\{W, J\}$
  - Replace the period of the switch  $W$  as  $L$

# Digit-Serial Adder (3/4)



# Digit-Serial Adder (4/4)

